

## ENGR-2300

## Electronic Instrumentation

## Quiz 1

Spring 2020

Name SOLUTIONSSection     

Question I (20 points) \_\_\_\_\_

Question II (20 points) \_\_\_\_\_

Question III (20 points) \_\_\_\_\_

Question IV (20 points) \_\_\_\_\_

LMS Question (20 points) (graded on LMS)

Total (80 points) \_\_\_\_\_

On all questions: SHOW ALL WORK. BEGIN WITH FORMULAS, THEN SUBSTITUTE VALUES AND UNITS. No credit will be given for numbers that appear without justification. Unless otherwise stated in a problem, provide 3 significant digits in answers. Read the entire quiz before answering any questions. Also it may be easier to answer parts of questions out of order.

Standard Resistor Values ( $\pm 5\%$ )						
1.0	10	100	1.0K	10K	100K	1.0M
1.1	11	110	1.1K	11K	110K	1.1M
1.2	12	120	1.2K	12K	120K	1.2M
1.3	13	130	1.3K	13K	130K	1.3M
1.5	15	150	1.5K	15K	150K	1.5M
1.6	16	160	1.6K	16K	160K	1.6M
1.8	18	180	1.8K	18K	180K	1.8M
2.0	20	200	2.0K	20K	200K	2.0M
2.2	22	220	2.2K	22K	220K	2.2M
2.4	24	240	2.4K	24K	240K	2.4M
2.7	27	270	2.7K	27K	270K	2.7M
3.0	30	300	3.0K	30K	300K	3.0M
3.3	33	330	3.3K	33K	330K	3.3M
3.6	36	360	3.6K	36K	360K	3.6M
3.9	39	390	3.9K	39K	390K	3.9M
4.3	43	430	4.3K	43K	430K	4.3M
4.7	47	470	4.7K	47K	470K	4.7M
5.1	51	510	5.1K	51K	510K	5.1M
5.6	56	560	5.6K	56K	560K	5.6M
6.2	62	620	6.2K	62K	620K	6.2M
6.8	68	680	6.8K	68K	680K	6.8M
7.5	75	750	7.5K	75K	750K	7.5M
8.2	82	820	8.2K	82K	820K	8.2M
9.1	91	910	9.1K	91K	910K	9.1M

Type	$R_{int}$ ( $\Omega$ )	$V_{oc}$ (V)	Capacity <sup>a</sup> continuous, to 1V/cell				Size (in)	Weight (gm)	Connec <sup>b</sup>	Comments
			(mAh) @ (mA)	(mAh) @ (mA)	(mAh) @ (mA)	(mAh) @ (mA)				
<b>9V "1604"</b>										
Le Clanche	35	9	300	1	160	10	0.65x1x1.9	35	S	
Heavy Duty	35	9	400	1	180	10	"	40	S	
Alkaline	2	9	500	1	470	10	"	55	S	280mAh@100mA
Lithium	18	9	1000	25	950	80	"	38	S	Kodak Li-MnO <sub>2</sub>

**I. Voltage Dividers and Battery Characteristics (20 points)**

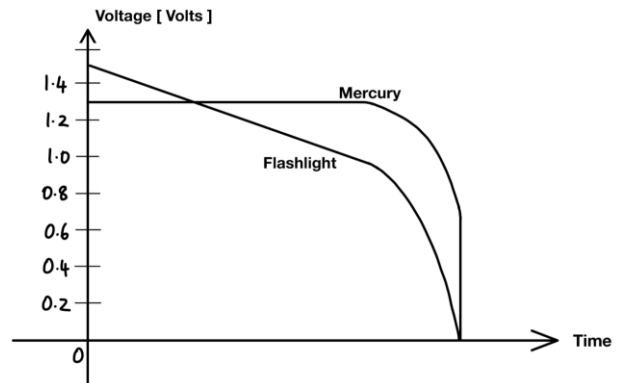
As stated on the cover page: Round answers to 3 significant digits. Show formulas first and show your work. No credit will be given for numbers that appear without justification.

a. The voltage-time curves for two types of batteries, flashlight and Mercury, are shown below.

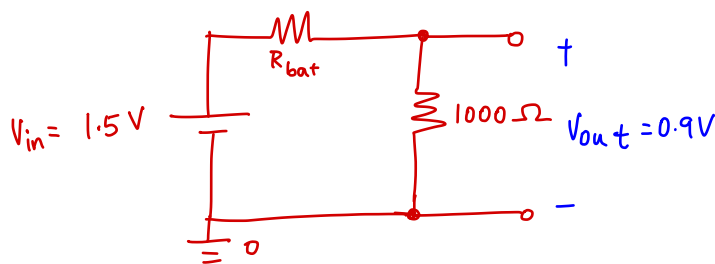
- i. Interpret the behavior in terms of the respective internal resistances. For this, you need to specifically comment about how their internal resistances change over time. (2pts)

Initially, the internal resistance of flashlight battery increases rapidly - at a much faster rate compared to that of the Mercury battery.

As time increases, just before the batteries are completely dead, the internal resistance of Mercury battery increases sharply.



- ii. A voltmeter whose resistance is  $1000\ \Omega$  measures the voltage of a worn-out  $1.5\text{V}$  flashlight battery (same one used in the previous part) as  $0.9\text{V}$ . What is the internal resistance of the battery? (3pts)

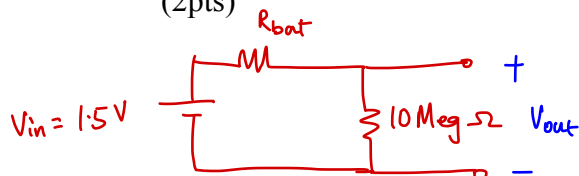


$$V_{out} = V_{in} \left( \frac{1000}{1000 + R_{bat}} \right)$$

$$\Rightarrow 900 + 0.9 R_{bat} = 1500$$

$$\Rightarrow R_{bat} = 666.67\ \Omega$$

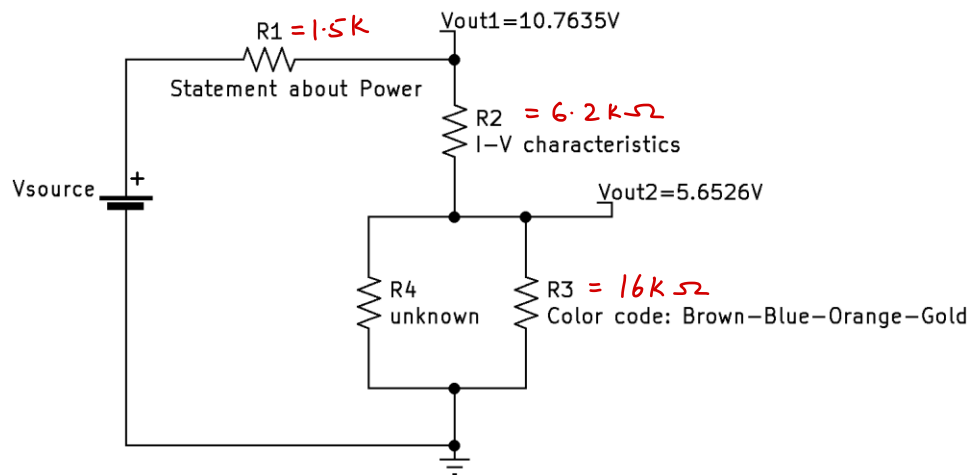
- iii. If the flashlight battery from the previous part had been measured using a vacuum tube voltmeter with a resistance of  $10\text{ Meg}\Omega$ , what voltage would have been read? (2pts)



$$R_{bat} \ll 10\text{ Meg}\Omega$$

$$\Rightarrow V_{out} \approx V_{in} = 1.5\text{V}$$

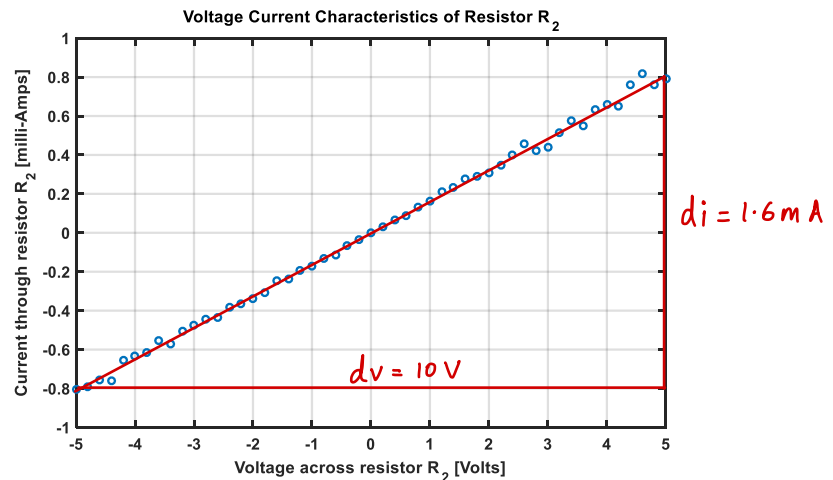
- b. Consider the circuit shown below. The DC voltage of the source,  $V_{source}$ , is unknown. We do know a few details about the resistors that are used in the circuit and the voltages  $V_{out1}$  and  $V_{out2}$  measured at points shown in the circuit to be 10.7635V and 5.6526V respectively.



- i. The power absorbed by resistor  $R_1$  is 4.3 mW when 1.693 mA flows through it. What is the value of  $R_1$ ? (1pt) Use the closest standard resistor value; refer to information on "Standard Resistor Values" at the beginning of the quiz.

$$R_1 = \frac{P_1}{I_1^2} = 1.5 \text{ k}\Omega$$

- ii. The voltage-current characteristics of resistor  $R_2$  is provided to the right. Determine the value of  $R_2$ . (2pts) Use the closest standard resistor value; refer to information on "Standard Resistor Values" at the beginning of the quiz.



$$R_2 = \frac{dv}{di} = \frac{10}{1.6\text{m}} = 6.25 \text{ k}\Omega$$

USE  $6.2 \text{ k}\Omega$  as standard resistor

- iii. The color code of resistor  $R_3$  is Brown-Blue-Orange-Gold. What is the value of resistor  $R_3$ ? (1pt)

$$R_3 = 16 \text{ k}\Omega \pm 5\%$$

- iv. Given  $V_{out1}=10.7635V$  and  $V_{out2} = 5.6526 V$  and the answers from the previous questions, determine the value of  $V_{source}$ ? (4pts) Hint: First find current through  $R_2$ .

$$\text{Voltage across } R_2 = V_{out1} - V_{out2} = 5.111 V$$

$$\text{Current through } R_2 = \frac{5.111}{R_2} = 0.824 \text{ mA} \text{ or } 824.4 \mu\text{A}$$

This is the same current through  $R_1$

$$\Rightarrow \text{Voltage across } R_1 = (1.5 \text{ k}\Omega)(0.824 \text{ mA}) = 1.237 V$$

$$V_{source} = V_{R1} + V_{out1} = 12 V$$

$$V_{source} = \boxed{12 V}$$

$$\text{or } = V_{R1} + V_{R2} + V_{out2} = 12 V$$

- v. Using all the information you have so far, determine how much current flows through resistor  $R_4$ ? (5pts)

$$\text{Current through } R_3 = \frac{V_{out2}}{R_3} = 0.3533 \text{ mA} \text{ or } 353.3 \mu\text{A}$$

$$\text{Current through } R_4 = \text{Current through } R_2 - \text{Current through } R_3$$

$$= 824.4 \mu\text{A} - 353.3 \mu\text{A}$$

$$= 471.1 \mu\text{A}$$

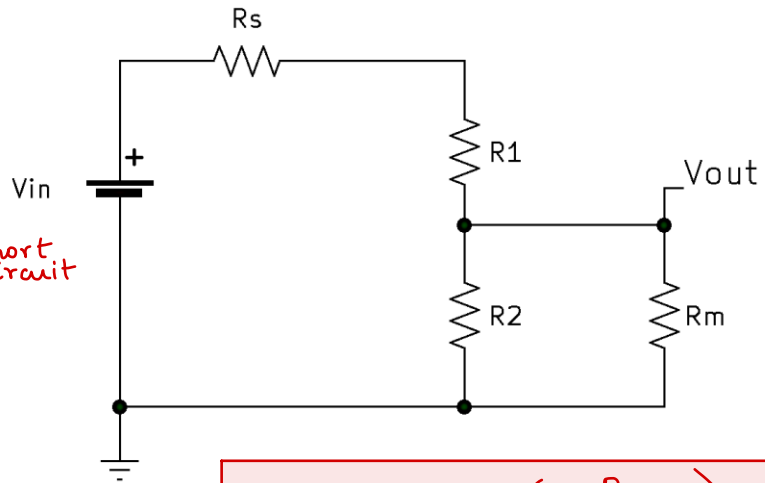
$$\text{Optional: } R_4 = \frac{V_{out2}}{I_{R4}} = 12 \text{ k}\Omega$$

$$\text{Current through } R_4 = \boxed{0.471 \text{ mA} \text{ or } 471.1 \mu\text{A}}$$

**II. Design Problem on Minimizing Measurement Error (20 points)**

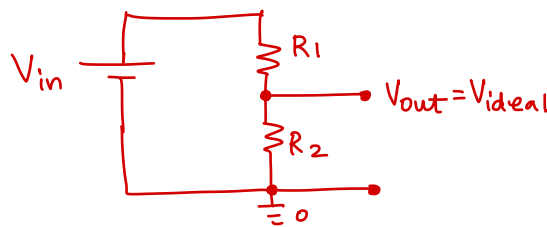
A voltage divider is connected to a source and a voltmeter as shown below. Ideally,  $R_s = 0$  and  $R_m = \infty$ . However, for one practical circuit,  $R_s \leq 125\Omega$  and  $R_m \geq 10k\Omega$ . The overall goal of this problem is to select  $R_1$  and  $R_2$  to minimize the error introduced by  $R_s$  and  $R_m$  when it is desired that  $\frac{V_{out}}{V_{in}} = 0.75$ . For simplicity use  $a = \frac{R_2}{R_1+R_2}$

- i. Determine an expression for  $V_{out}$  in terms of  $V_{in}$ ,  $R_1$ , and  $R_2$  under ideal conditions, i.e.  $R_s = 0$  and  $R_m = \infty$ . Denote this output voltage as  $V_{ideal}$ . (3pts)



open circuit

short circuit



$$V_{ideal} = V_{in} \left( \frac{R_2}{R_1 + R_2} \right)$$

or  $= a V_{in}$ , where  $a = \frac{R_2}{R_1 + R_2}$

- ii. Determine an expression for  $V_{out}$  in terms of  $V_{in}$ ,  $R_1$ ,  $R_2$ ,  $R_s$ , and  $R_m$  under practical conditions. Denote this output voltage as  $V_{pract}$ . (4pts)

Let  $R_p = R_2 \parallel R_m = \frac{R_2 R_m}{R_2 + R_m}$

$$V_{out} = V_{pract} = V_{in} \left( \frac{R_p}{R_p + R_1 + R_s} \right) = V_{in} \left[ \frac{R_2 R_m}{R_2 R_m + (R_2 + R_m)(R_1 + R_s)} \right]$$

Or in terms of  $a \rightarrow$

$$V_{pract} = V_{in} \left[ \frac{a R_2 R_m}{(a R_s + R_2)(R_m + R_2) - a R_2^2} \right]$$

- iii. Find an expression for measurement error 'e'. Define  $e = \frac{V_{ideal} - V_{pract}}{V_{ideal}}$ . Simplify such that there are no fractions in the numerator or denominator. (5pts)

$$e = \frac{V_{ideal} - V_{pract}}{V_{ideal}} = \frac{aV_{in} - V_{pract}}{aV_{in}} = 1 - \frac{V_{pract}}{aV_{in}}$$

$$= 1 - \frac{R_2 R_m}{(aR_s + R_2)(R_m + R_2) - aR_2^2}$$

OR

$$e = 1 - \frac{V_{pract}}{V_{ideal}} = 1 - \frac{R_m(R_1 + R_2)}{R_2 R_m + (R_2 + R_m)(R_1 + R_s)}$$

- iv. Describe how you would analytically minimize this error. You just need to describe your steps; not actually go through them. (4pts)

We can use calculus and set  $\frac{de}{dR_2} = 0$  to find the best value of  $R_2$ .

We can use spreadsheet computer program to find the best value of  $R_2$  that minimizes  $e$ .

- v. It turns out that the error is minimized when  $R_2 = \sqrt{3 \times R_m \times R_s}$ . To obtain the desired  $\frac{V_{out}}{V_{in}} = 0.75$  and using the worst case values of  $R_m$  and  $R_s$  provided in the problem statement, find the values of  $R_1$  and  $R_2$  that will satisfy this design problem. (4pts)

Worst case:  $R_m = 10k\Omega$  and  $R_s = 125\Omega$

$$\Rightarrow R_2 = \sqrt{3 \times 10k \times 125} = 1936.5 \Omega$$

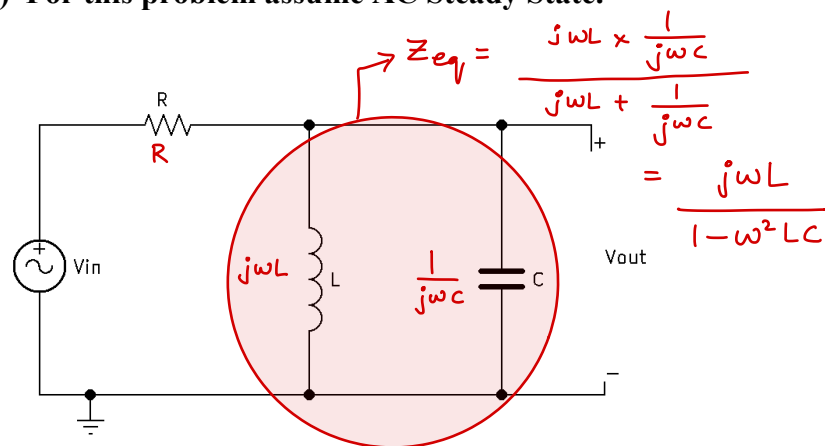
$$V_{out} = a V_{in} \Rightarrow a = \frac{V_{out}}{V_{in}} = 0.75$$

$$a = \frac{R_2}{R_1 + R_2} \Rightarrow R_1 = \frac{R_2(1-a)}{a} = 645.5 \Omega$$



### III. Filters & Transfer Functions (20 points) For this problem assume AC Steady State.

- a. Find the transfer function of the circuit shown. Simplify such that there are no fractions in the numerator or denominator of the transfer function.  $H(j\omega) = V_{out}(j\omega)/V_{in}(j\omega)$  (4pts)



$$H(j\omega) = \frac{V_{out}}{V_{in}} = \frac{Z_{eq}}{Z_{eq} + R}$$

$$= \frac{\frac{j\omega L}{1 - \omega^2 LC}}{\frac{j\omega L}{1 - \omega^2 LC} + R} = \frac{j\omega L}{R(1 - \omega^2 LC) + j\omega L}$$

- b. Determine the magnitude and phase of the transfer function for the circuit for very small frequency and for very high frequency. Do not take this to 0 or infinite Hz. (4pts)

Low frequency

$$|H(j\omega)| = \frac{j\omega L}{R}$$

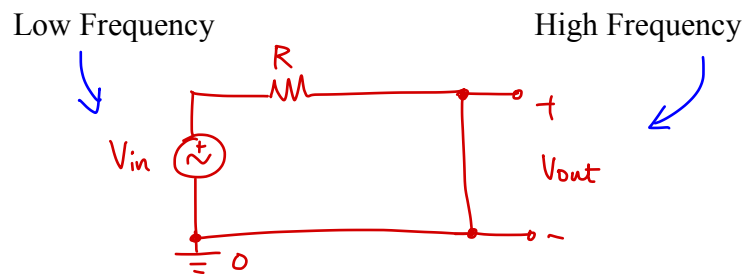
$$\angle H(j\omega) = \angle \frac{j\omega L}{R} = 90^\circ$$

High frequency

$$|H(j\omega)| = \frac{j\omega L}{-\omega^2 R L C} = \frac{-j}{\omega R C}$$

$$\angle H(j\omega) = \angle \frac{-j}{\omega R C} = -90^\circ$$

- c. **Redraw the circuit** and simplify the circuit for operation at low and high frequency. For this part you take it to extremes, low frequency is dc operation. High is approaching infinity. (2pts)



Same circuit for both

- d. Corner frequency of a filter,  $f_c$ , is defined as the frequency at which the magnitude of the transfer function,  $|H(j\omega)|$ , is 0.707, or  $\frac{1}{\sqrt{2}}$ . Alternatively, this can be expressed as the frequency at which the power at the output is half of the power at the input. **Derive the corner frequency or frequencies** for this circuit when  $R=10\ \Omega$ ,  $L=2\text{H}$ , and  $C=0.5\text{F}$ ? (5 pts)

$$H(j\omega) = \frac{j\omega L}{R(1-\omega^2 LC) + j\omega L} \Rightarrow |H(j\omega)| = \frac{|j\omega L|}{|R(1-\omega^2 LC) + j\omega L|}$$

$$= \frac{\omega L}{\sqrt{R^2(1-\omega^2 LC)^2 + \omega^2 L^2}} \stackrel{\text{set}}{=} \frac{1}{\sqrt{2}}$$

$$= \frac{2\omega_c}{\sqrt{100(1-\omega_c^2)^2 + 4\omega_c^2}} = \frac{1}{\sqrt{2}}$$

$$8\omega_c^2 = 100(1-\omega_c^2)^2 + 4\omega_c^2 \Rightarrow 100\omega_c^4 - 204\omega_c^2 + 100 = 0$$

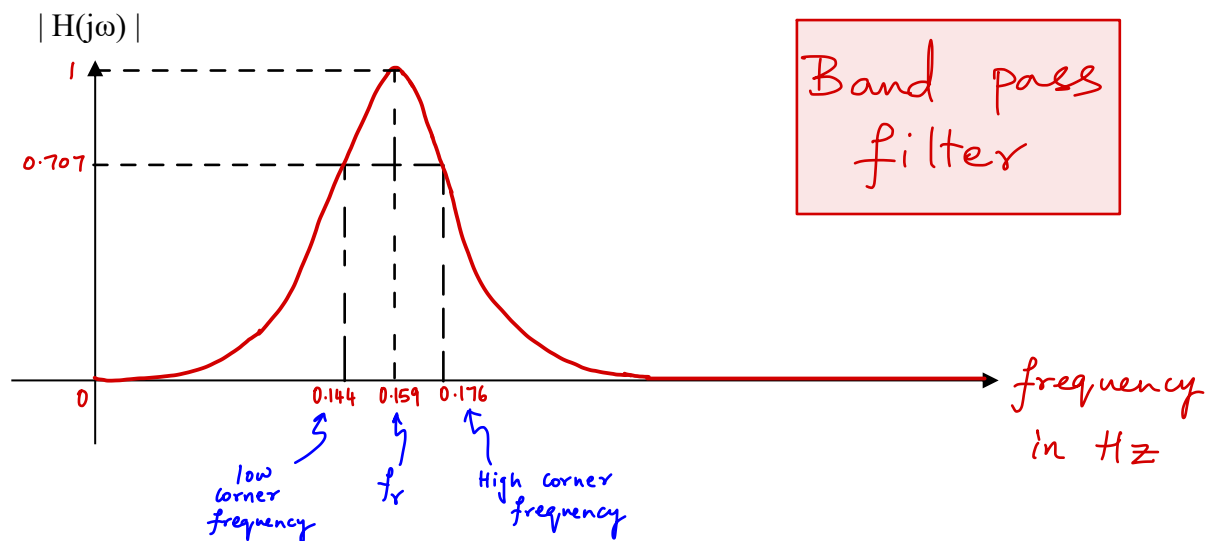
$$\omega_c^2 = \frac{204 \pm \sqrt{204^2 - 4 \times 10^4}}{200} = 1.221, 0.819$$

$$\Rightarrow \omega_c = 1.1050, 0.9050 \Rightarrow f_c = \frac{\omega_c}{2\pi} = 0.176\text{ Hz and } 0.144\text{ Hz}$$

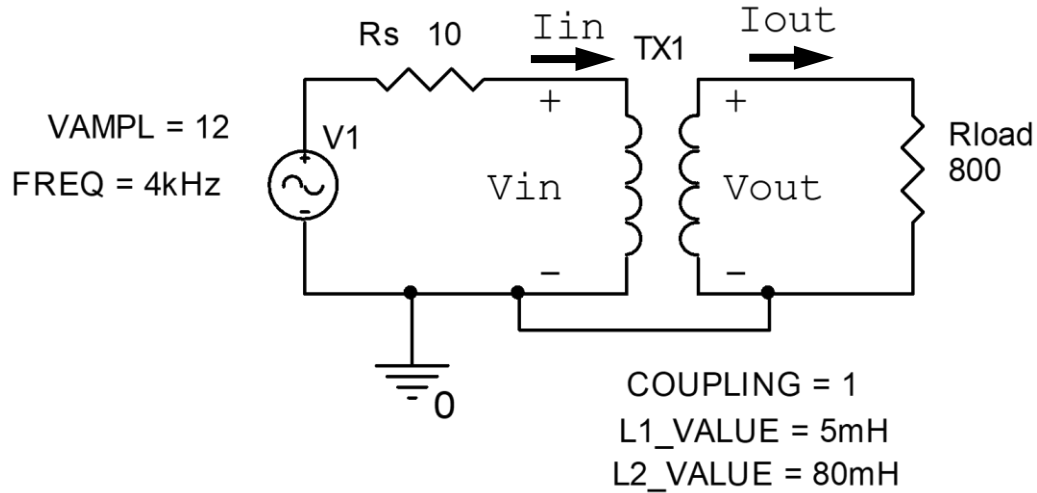
- e. Resonant frequency of a filter,  $f_r$ , is defined as the frequency at which the magnitude of the transfer function,  $|H(j\omega)|$ , is at a maximum or minimum, depending on the type of filter. Determine the resonant frequency of the circuit,  $f_r$  using the same R, L, and C values as the previous part and the equation from the crib sheet. (2 pts)

$$f_r = \frac{1}{2\pi\sqrt{LC}} = 0.159 \text{ Hz}$$

- f. Provide a rough sketch for the AC sweep analysis ( $|H(j\omega)|$  vs. frequency) for this filter. Clearly mark the corner frequency and resonant frequency. Label the x-axis. **What type of filter** have we been working with for this problem? (3pts)



## IV – Phasors and Transformers (20 points)



a. Assume an ideal transformer with full coupling.

- i. For the given information, determine the turns ratio,  $a$ . And determine the ratios  $V_{out}/V_{in}$ ,  $I_{out}/I_{in}$  and the transformer input impedance  $R_{in}$ . ( $R_{in}$  is  $V_{in}/I_{in}$ ) (6 pts)

$$a = \sqrt{\frac{L_2}{L_1}} = \sqrt{\frac{80}{5}} = 4$$

$$\frac{V_{out}}{V_{in}} = a = 4$$

$$\frac{I_{out}}{I_{in}} = \frac{1}{a} = \frac{1}{4}$$

$$R_{in} = \frac{R_{Load}}{a^2} = 50 \Omega$$

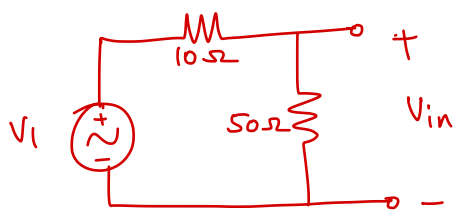
$$a = \underline{4} \quad (2 \text{ pts})$$

$$V_{out}/V_{in} = \underline{4} \quad (1 \text{ pt})$$

$$I_{out}/I_{in} = \underline{0.25} \quad (1 \text{ pt})$$

$$R_{in} = \underline{50 \Omega} \quad (2 \text{ pts})$$

- ii. **Solve for  $V_{in}$**  (voltage across the input terminals of the ideal transformer) **and  $V_{out}$** , the voltage across the output terminals and the of the ideal transformer. **Assume the phase of  $V_1$  is zero degrees and give the answer in the form of  $v(t)=V_1\cos(\omega t+\theta_1)$**  (3 pts)



$$|V_{in}| = \frac{50}{50+10} \times 12 = 10V$$

$$\angle V_{in} = \angle V_1 = 0^\circ$$

$$\omega = 2\pi \times 4 \times 10^3 = 8000\pi \text{ rad/s}$$

$$V_{out} = 4 \times V_{in}$$

$$V_{in}(t) = \underline{10 \cos(8000\pi t + 0^\circ)} \text{ V}$$

$$V_{out}(t) = \underline{40 \cos(8000\pi t + 0^\circ)} \text{ V}$$

- iii. Above you were told to assume that the transformer is ideal. For that to be valid, the impedance of the primary inductor should be much larger than the source resistance. Is that valid in this case? Explain or justify. Would it be valid at if the signal source was at 60Hz? (3pts)

@ 4 kHz  $|j\omega L_1| = 8000\pi \times 5 \times 10^{-3} = 125.66 \Omega$

2pts

$125.66 \Omega \gg 10 \Omega \Rightarrow$  Reasonably true,  
So **yes** valid.

@ 60 Hz  $|j\omega L_1| = 2\pi \times 60 \times 5 \times 10^{-3} = 1.885 \Omega$

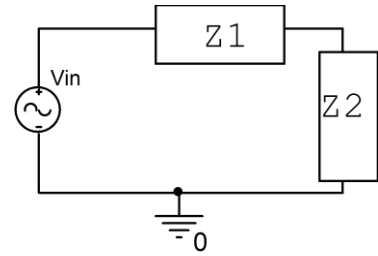
1pt

$1.885 \Omega$  is much less than  $10 \Omega$

**No**

- b. Phasors: This circuit shown has 2 complex impedances,  $Z_1$  and  $Z_2$ , connected as shown.

Given:  $V_{in} = 5V \angle 45^\circ$  and the voltage across  $Z_2$  is measured to be  $V_{z2} = 3V \angle 30^\circ$



- i. Write  $V_{in}$  and  $V_{z2}$  in Cartesian form. (2pts)

$$V_{in} = 5 \cos(45^\circ) + j 5 \sin(45^\circ)$$

$$= \boxed{3.54 + j 3.54}$$

$$V_{z2} = 3 \cos(30^\circ) + j 3 \sin(30^\circ)$$

$$= \boxed{2.6 + j 1.5}$$

- ii. Determine  $V_{z1}$ , the voltage across  $Z_1$  in Cartesian and polar form (3pts)

$$V_{z1} = V_{in} - V_{z2} = \boxed{0.9374 + j 2.04} \quad \text{Cartesian}$$

$$= \boxed{2.241 \angle 65.32^\circ} \quad \text{polar}$$

- iii. If  $Z_2$  is a  $1k\Omega$  resistor, and only a resistor, what is the current through  $Z_2$  in both polar and Cartesian form? (3pts)

$$I_{z2} = \frac{V_{z2}}{Z_2} = \frac{3 \angle 30^\circ}{1k} = \boxed{3 \angle 30^\circ \text{ mA}} \quad 1pt$$

$$= \frac{2.6 + j 1.5}{1k} = \boxed{(2.6 + j 1.5) \text{ mA}} \quad 2pt$$